NON-LINEAR CONVOLUTION: A NEW APPROACH FOR THE AURALIZATION OF DISTORTING SYSTEMS

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Goals for Auralization

- Transform the results of objective electroacoustics measurements to audible sound samples suitable for listening tests
- Traditional auralization is based on linear convolution: this does not replicates faithfully the nonlinear behaviour of most transducers
- The new method presented here overcomes to this strong limitation, providing a simplified treatment of memory-less distortion

Methods

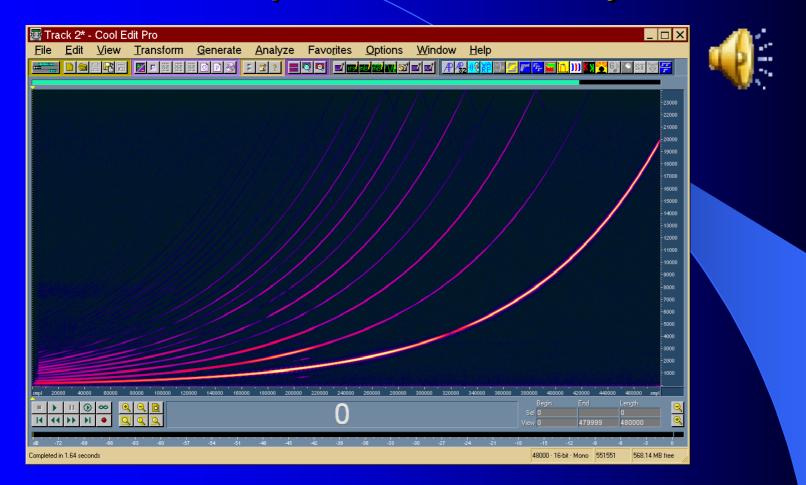
- We start from a measurement of the system based on exponential sine sweep (Farina, 108th AES, Paris 2000)
- Diagonal Volterra kernels are obtained by postprocessing the measurement results
- These kernels are employed as FIR filters in a multiple-order convolution process (original signal, its square, its cube, and so on are convolved separately and the result is summed)

Exponential sweep measurement



 The excitation signal is a sine sweep with constant amplitude and exponentially-increasing frequency

Raw response of the system



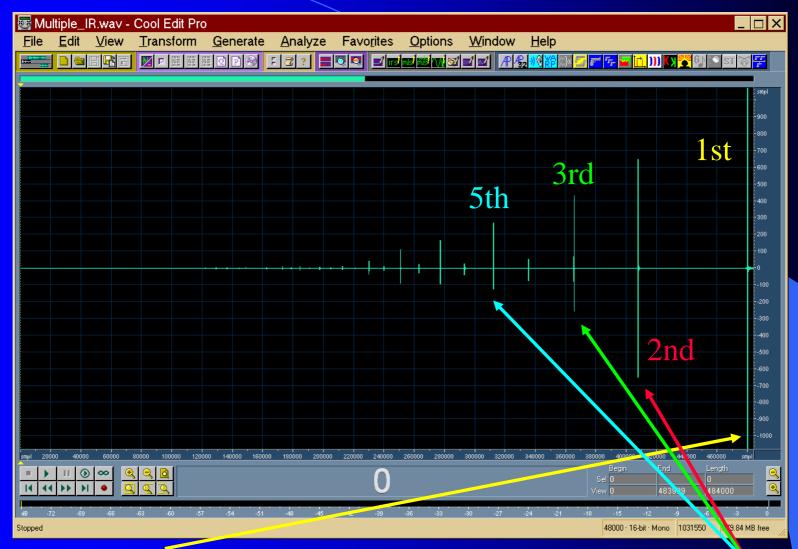
Many harmonic orders do appear as colour stripes

Deconvolution of system's impulse response



The deconvolution is obtained by convolving the raw response with a suitable inverse filter

Multiple impulse response obtained



The last peak is the linear impulse response, the preceding ones are the harmonic distortion orders

Auralization by linear convolution



Convolving a suitable sound sample with the linear IR, the frequency response and temporal transient effects of the system can be simulated properly

What's missing in linear convolution ?

- No harmonic distortion, nor other nonlinear effects are being reproduced.
- From a perceptual point of view, the sound is judged "cold" and "innatural"
- A comparative test between a strongly nonlinear device and an almost linear one does not reveal any audible difference, because the nonlinear behavior is removed for both

Theory of nonlinear convolution

- The basic approach is to convolve separately, and then add the result, the linear IR, the second order IR, the third order IR, and so on.
- Each order IR is convolved with the input signal raised at the corresponding power:

$$y(n) = \sum_{i=0}^{M-1} h_1(i) \cdot x(n-i) + \sum_{i=0}^{M-1} h_2(i) \cdot x^2(n-i) + \sum_{i=0}^{M-1} h_3(i) \cdot x^3(n-i) + \dots$$

The problem is that the required multiple IRs **are not** the results of the measurements: they are instead the diagonal terms of Volterra kernels

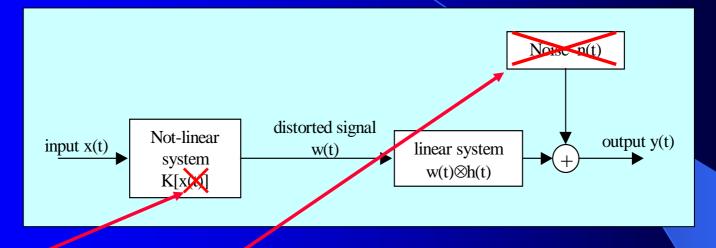
Volterra kernels and simplification

• The general Volterra series expansion is defined as:

$$\begin{aligned} \mathbf{y}(\mathbf{n}) &= \sum_{i_1=0}^{M-1} h_1(i_1) \cdot \mathbf{x}(\mathbf{n}-i_1) + \sum_{i_1=0}^{M-1} \sum_{i_2=0}^{M-1} h_2(i_1,i_2) \cdot \mathbf{x}(\mathbf{n}-i_1) \cdot \mathbf{x}(\mathbf{n}-i_2) + \\ &+ \sum_{i_1=0}^{M-1} \sum_{i_2=0}^{M-1} \sum_{i_3=0}^{M-1} h_3(i_1,i_2,i_3) \cdot \mathbf{x}(\mathbf{n}-i_1) \cdot \mathbf{x}(\mathbf{n}-i_2) \cdot \mathbf{x}(\mathbf{n}-i_3) + \dots \end{aligned}$$

This explains also nonlinear effect with memory, as the system output contains also products of previous sample values with different delays

Memoryless distortion followed by a linear system with memory



- The first nonlinear system is assumed to be memory-less, so only the diagonal terms of the Volterra kernels need to be taken into account.
- Furthermore, we neglect the noise, which is efficiently rejected by the sine sweep measurement method.

Volterra kernels from the measurement results The measured multiple IRs h' can be defined as: $y(t) = h'_1 \otimes \sin[\omega_{var}] + h'_2 \otimes \sin[2 \cdot \omega_{var}] + h'_3 \otimes \sin[3 \cdot \omega_{var}] + \dots$

We need to relate them to the simplified Volterra kernels h:

$$\mathbf{y}(t) = \mathbf{h}_1 \otimes \sin[\omega_{\text{var}}] + \mathbf{h}_2 \otimes \sin^2[\omega_{\text{var}}] + \mathbf{h}_3 \otimes \sin^3[\omega_{\text{var}}] + \dots$$

Trigonometry can be used to expand the powers of the sinusoidal terms:

$$\sin^{2}(\omega \cdot \tau) = \frac{1}{2} - \frac{1}{2} \cdot \cos(2 \cdot \omega \cdot \tau) \qquad \sin^{3}(\omega \cdot \tau) = \frac{3}{4} \cdot \sin(\omega \cdot \tau) - \frac{1}{4} \cdot \sin(3 \cdot \omega \cdot \tau)$$
$$\sin^{4}(\omega \cdot \tau) = \frac{3}{8} - \frac{1}{2} \cdot \cos(2 \cdot \omega \cdot \tau) + \frac{1}{8} \cdot \cos(4 \cdot \omega \cdot \tau)$$
$$\sin^{5}(\omega \cdot \tau) = \frac{5}{8} \cdot \sin(\omega \cdot \tau) - \frac{5}{16} \cdot \sin(3 \cdot \omega \cdot \tau) + \frac{1}{16} \cdot \sin(5 \cdot \omega \cdot \tau)$$

Finding the connection point

Going to frequency domain by taking the FFT, the first equation becomes:

$$Y(\omega) = \overline{H'_1}[\omega] \cdot X[\omega] + \overline{H'_2}[\omega] \cdot X[\omega/2] + \overline{H'_3}[\omega] \cdot X[\omega/3] + \dots$$

Doing the same in the second equation, and substituting the trigonometric expressions for power of sines, we get:

$$Y(\omega) = \overline{H_1} + \frac{3}{4} \cdot \overline{H_3} + \frac{5}{8} \cdot \overline{H_5} \cdot X[\omega] + \left[-\frac{1}{2} \cdot \overline{H_2} - \frac{1}{2} \cdot \overline{H_4} \right] j \cdot X[\omega/2] + \left[-\frac{1}{4} \cdot \overline{H_3} - \frac{5}{16} \cdot \overline{H_5} \right] \cdot X[\omega/3] + \frac{1}{8} \cdot \overline{H_4} \cdot j \cdot X[\omega/4] + \frac{1}{16} \cdot \overline{H_5} \cdot X[\omega/5] + \dots$$

The terms in square brackets have to be equal to the corresponding measured transfer functions H' of the first equation

Solution

• Thus we obtain a linear equation system:

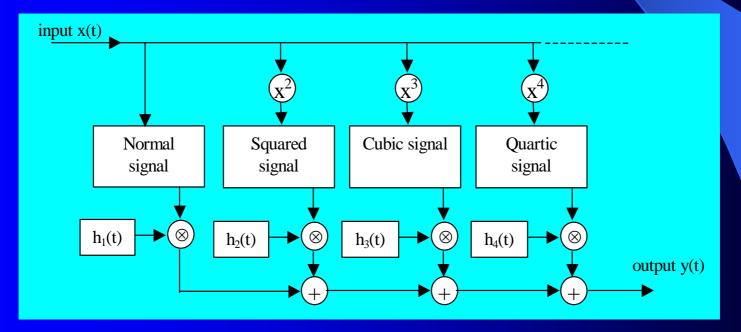
$$\begin{cases} \overline{H'}_1 = \overline{H}_1 + \frac{3}{4} \cdot \overline{H}_3 + \frac{5}{8} \cdot \overline{H}_5 \\ \overline{H'}_2 = -j \cdot \frac{1}{2} \cdot \left[\overline{H}_2 + \overline{H}_4 \right] \\ \overline{H'}_3 = -\frac{1}{4} \cdot \overline{H}_3 - \frac{5}{16} \cdot \overline{H}_5 \\ \overline{H'}_4 = j \cdot \frac{1}{8} \cdot \overline{H}_4 \\ \overline{H'}_5 = \frac{1}{16} \cdot \overline{H}_5 \end{cases}$$

We can easily solve it, obtaining the required Volterra kernels as a function of the measured multiple-order IRs:

$$\begin{cases} \overline{H}_{1} = \overline{H'}_{1} + 3 \cdot \overline{H'}_{3} + 5 \cdot \overline{H'}_{5} \\ \overline{H}_{2} = 2 \cdot j \cdot \overline{H'}_{2} + 8 \cdot j \cdot \overline{H'}_{4} \\ \overline{H}_{3} = -4 \cdot \overline{H'}_{3} - 20 \cdot \overline{H'}_{5} \\ \overline{H}_{4} = -8 \cdot j \cdot \overline{H'}_{4} \\ \overline{H}_{5} = 16 \cdot \overline{H'}_{5} \end{cases}$$

Non-linear convolution

As we have got the Volterra kernels already in frequency domain, we can efficiently use them in a multiple convolution algorithm implemented by overlap-and-save of the partitioned input signal:



Software implementation

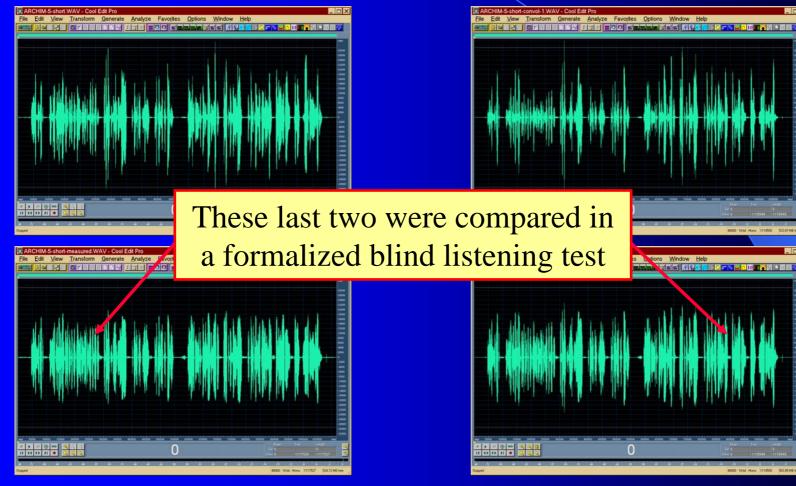
Although today the algorithm is working off-line (as a mix of manual CoolEdit operations and some Matlab processing), a more efficient implementation as a CoolEdit plugin is being worked out:

| Multiple Convolution with Clipboard | | | | | |
|-------------------------------------|--------|--------------------------------------|--|--|--|
| Stimulus | | Impulse Response | | | |
| N. of sweeps / measurement | 4 | N. of samples for each response 4096 | | | |
| Start Frequency (Hz) | 40 | N. of first samples to skip 0 | | | |
| End Frequency (Hz) | 18000 | Number of Harmonic Orders 10 | | | |
| Sweep duration (s or samples) | 10 | Autoscale and remove DC component | | | |
| Silence duration (s or samples) | 1 | User: Angelo Farina | | | |
| OK <u>H</u> elp | Cancel | Reg. key: | | | |

This will allow for real-time operation even with a very large number of filter coefficients

Audible evaluation of the performance

Original signal



Live recording



Non-linear multi convolution

Linear convolution

Subjective listening test

- A/B comparison
- Live recording & non-linear auralization
- 12 selected subjects
- 4 music samples
- 9 questions
- 5-dots horizontal scale
- Simple statistical analysis of the results
- A was the live recording, B was the auralization, but the listener did not know this

95% confidence intervals of the responses

| 📲 Risposte soggettive | e _ 🗆 🗙 | | | |
|--|------------------------------|--|--|--|
| Brano n. 1 2 3 | 4 A B ► □ | | | |
| D:\Convol_altop_lamiera\05RebeccaPidgeon-porta.WAV | | | | |
| -Domanda 1 | | | | |
| A & B are identical | A & B are quite different | | | |
| -Domanda 2 | | | | |
| A is more enveloping | C C C C B is more enveloping | | | |
| - Domanda 3 | | | | |
| A has better timber | C + B has better timber | | | |
| -Domanda 4 | | | | |
| A is more dry | C B is more dry | | | |
| Domanda 5 | | | | |
| A is more distorted | B is more distorted | | | |
| -Domanda 6 | | | | |
| A has more treble | C C 🔶 C 🔶 B has more treble | | | |
| -Domanda 7 | | | | |
| A has more medium | O C B has more medium | | | |
| -Domanda 8 | | | | |
| A has more bass | C B has more bass | | | |
| Domanda 9 | | | | |
| A is more pleasant | C B is more pleasant | | | |
| Precedente | Successivo Fine | | | |

Conclusion

Statistical parameters – more advanced statistical methods would be advisable for getting more significant results

| Question Number | Average score | 2.67 * Std. Dev. |
|-------------------------|---------------|------------------|
| 1 (identical-different) | 1.25 | 0.76 |
| 3 (better timber) | 3.45 | 1.96 |
| 5 (more distorted) | 2.05 | 1.34 |
| 9 (more pleasant) | 3.30 | 2.16 |

Final remarks

- The CoolEdit plugin is planned to be released in two months it will be downloadable from <u>HTTP://www.ramsete.com/aurora</u>
- The sound samples employed for the subjective test are available for download at <u>HTTP://pcangelo.eng.unipr.it/public/AES110</u>
- The new method will be employed for realistic reproduction in a listening room of the behaviour of car sound systems